## Self Protected Very Low Iq **High Side Driver with Analog Current Sense**

## NCV84140

The NCV84140 is a fully protected single channel high side driver that can be used to switch a wide variety of loads, such as bulbs, solenoids, and other actuators. The device incorporates advanced protection features such as active inrush current management, over-temperature shutdown with automatic restart and an overvoltage active clamp. A dedicated Current Sense pin provides precision analog current monitoring of the output as well as fault indication of short to V<sub>D</sub>, short circuit to ground and OFF state open load detection. An active high Current Sense Enable pin allows all diagnostic and current sense features to be enabled.

#### Features

- Short Circuit Protection with Inrush Current Management
- CMOS (3 V / 5 V) Compatible Control Input
- Very Low Standby Current
- Very Low Current Sense Leakage
- Proportional Load Current Sense
- Current Sense Enable
- Off State Open Load Detection
- Output Short to V<sub>D</sub> Detection
- Overload and Short to Ground Indication
- Thermal Shutdown with Automatic Restart
- Undervoltage Shutdown
- Integrated Clamp for Inductive Switching
- Loss of Ground and Loss of V<sub>D</sub> Protection
- ESD Protection
- Reverse Battery Protection with External Components
- AEC-Q100 Qualified
- This is a Pb–Free Device

#### **Typical Applications**

- Switch a Variety of Resistive, Inductive and Capacitive Loads
- Can Replace Electromechanical Relays and Discrete Circuits
- Automotive / Industrial

#### FEATURE SUMMARY

R <sub>DSon</sub> (typical) T <sub>J</sub> = 25°C	R <sub>ON</sub>	140	mΩ
Output Current Limit (typical)	I <sub>lim</sub>	12	А
OFF-state Supply Current (max)	I <sub>D(off)</sub>	0.5	μΑ

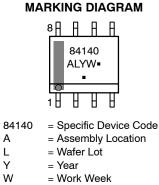


## **ON Semiconductor®**

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SOIC-8 CASE 751-07 STYLE 11

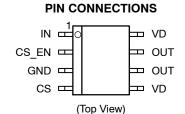


= Pb-Free Package

L

Y

(Note: Microdot may be in either location)



#### **ORDERING INFORMATION**

Device	Package	Shipping <sup>†</sup>
NCV84140DR2G	SOIC-8 (Pb-Free)	2500 / Tape & Reel

+For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specification Brochure, BRD8011/D.

### **BLOCK DIAGRAM & PIN CONFIGURATION**

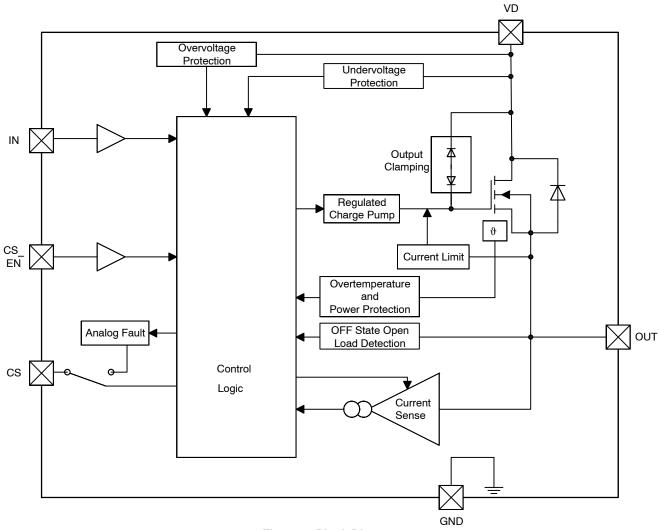


Figure 1. Block Diagram

#### Table 1. SO8 PACKAGE PIN DESCRIPTION

Pin #	Symbol	Description
1	IN	Logic Level Input
2	CS_EN	Current Sense Enable
3	GND	Ground
4	CS	Analog Current Sense Output
5	VD	Supply Voltage
6	OUT	Output
7	OUT	Output
8	VD	Supply Voltage

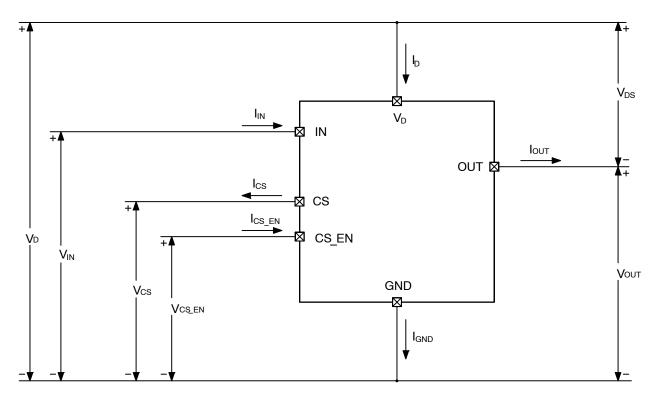


Figure 2. Voltage and Current Conventions

Connection	Input	Output	Current Sense	Current Sense Enable
Floating	Х	Х	Not Allowed	Х
To Ground	Through 10 k $\Omega$ resistor	Not Allowed	Through 1 k $\Omega$ Resistor	Through 10 k $\Omega$ resistor

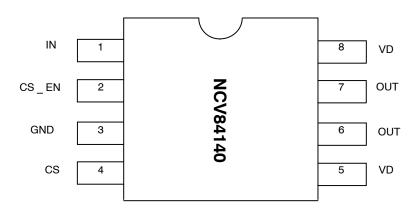


Figure 3. Pin Configuration (Top View)

#### **ELECTRICAL SPECIFICATIONS**

#### **Table 3. MAXIMUM RATINGS**

Rating	Symbol	١	/alue	Unit
DC Supply Voltage	V <sub>D</sub>	-0.3	38	V
Max Transient Supply Voltage (Note 1) Load Dump – Suppresses	U <sub>S</sub> *	-	45	V
Input Voltage	V <sub>IN</sub>	-10	10	V
Input Current	I <sub>IN</sub>	-5	5	mA
Reverse Ground Pin Current	I <sub>GND</sub>	-	-200	mA
Output Current (Note 2)	I <sub>OUT</sub>	-6	Internally Limited	А
Reverse CS Current	I <sub>CS</sub>	-	-200	mA
CS Voltage	V <sub>CS</sub>	V <sub>D</sub> – 41	V <sub>D</sub>	V
CS_EN Voltage	V <sub>CS_EN</sub>	-10	10	V
CS_EN Current	I <sub>CS_EN</sub>	-5	5	mA
Power Dissipation Tc = $25^{\circ}$ C (Note 5)	P <sub>tot</sub>		1.17	W
Electrostatic Discharge (Note 3) (HBM Model 100 pF / 1500 Ω) Input Current Sense Current Sense Enable Output V <sub>D</sub> Charged Device Model CDM-AEC-Q100-011	V <sub>ESD</sub>	4 4 4 4 750	- - - - -	DC kV kV kV kV kV V
Single Pulse Inductive Load Switching Energy (L = 5 mH, $I_L$ = 3.84 A, $T_{Jstart}$ = 150°C, $V_D$ tied to GND during inductive discharge)	E <sub>AS</sub>	-	36.8	mJ
Operating Junction Temperature	TJ	-40	+150	°C
Storage Temperature	T <sub>storage</sub>	-55	+150	°C

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected. 1. Load Dump Test B (with centralized load dump suppression) according to ISO16750–2 standard. Guaranteed by design. Not tested in

production. Passed Class C (or A, B) according to ISO16750–1. 2. Reverse Output current has to be limited by the load to stay within absolute maximum ratings and thermal performance.

3. This device series incorporates ESD protection and is tested by the following methods:

ESD Human Body Model tested per AEC-Q100-002 (JS-001-2017)

Field Induced Charge Device Model ESD characterization is not performed on plastic molded packages with body sizes smaller than 2 x 2 mm due to the inability of a small package body to acquire and retain enough charge to meet the minimum CDM discharge current waveform characteristic defined in JEDEC JS-002-2018

#### **Table 4. THERMAL RESISTANCE RATINGS**

Parameter	Symbol	Max. Value	Units
Thermal Resistance Junction-to-Lead (Note 4) Junction-to-Ambient (Note 4) Junction-to-Ambient (Note 5)	R <sub>θJL</sub> R <sub>θJA</sub> R <sub>θJA</sub>	29 65 106	°C/W

4. 645 mm<sup>2</sup> pad size, mounted on four-layer 1s2p PCB - FR4, 2 oz. Cu thickness for top layer and 1 oz. Cu thickness for inner layers (planes not electrically connected)

5. 2 cm<sup>2</sup> pad size, mounted on single-layer 1s0p PCB - FR4, 2 oz. Cu thickness

## $\label{eq:constraint} \textbf{ELECTRICAL CHARACTERISTICS} ~ (7~V \le V_D \le 28~V; ~-40^\circ C \le T_J \le 150^\circ C ~ \text{unless otherwise specified})$

#### Table 5. POWER

				Value		
Rating	Symbol	Conditions	Min	Тур	Мах	Unit
Operating Supply Voltage	VD		4	-	28	V
Undervoltage Shutdown	V <sub>UV</sub>		-	3.5	4	V
Undervoltage Shutdown Hysteresis	V <sub>UV_hyst</sub>		-	0.4	-	V
On Resistance	R <sub>ON</sub>	I <sub>OUT</sub> = 1 A, T <sub>J</sub> = 25°C	-	140	-	mΩ
		I <sub>OUT</sub> = 1 A, T <sub>J</sub> = 150°C	-	-	295	
		$I_{OUT}$ = 1 A, $V_D$ = 4.5 V, $T_J$ = 25°C	-	-	210	
Supply Current (Note 6)	۱ <sub>D</sub>	OFF-state: $V_D = 13 V$ , $V_{IN} = V_{OUT} = 0 V$ , $Tj = 25^{\circ}C$	-	0.2	0.5	μΑ
		OFF-state: $V_D = 13 V$ , $V_{IN} = V_{OUT} = 0 V$ , $Tj = 85^{\circ}C$	-	0.2	0.5	μΑ
		OFF-state: V <sub>D</sub> = 13 V, V <sub>IN</sub> = V <sub>OUT</sub> = 0 V, Tj = 125°C	-	-	3	μΑ
		ON-state: $V_D = 13 V$ , $V_{IN} = 5 V$ , $I_{OUT} = 0 A$	-	1.9	3.5	mA
On State Ground Current	I <sub>GND(ON)</sub>	$V_D$ = 13 V, $V_{CS\_EN}$ = 5 V $V_{IN}$ = 5 V, $I_{OUT}$ = 1 A	-	-	6	mA
Output Leakage Current	١L	$V_{IN} = V_{OUT} = 0 \text{ V}, V_D = 13 \text{ V}, Tj = 25^{\circ}\text{C}$	0	-	0.5	μA
		V <sub>IN</sub> = V <sub>OUT</sub> = 0 V, V <sub>D</sub> = 13 V, Tj = 125°C	0	-	3	1

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.6. Includes PowerMOS leakage current.

## Table 6. LOGIC INPUTS (V\_D = 13.5 V; $-40^\circ C \leq T_J \leq 150^\circ C)$

				Value		
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Input Voltage – Low	V <sub>IN_low</sub>		-	-	0.9	V
Input Current – Low	I <sub>IN_low</sub>	V <sub>IN</sub> = 0.9 V	1	-	-	μΑ
Input Voltage – High	V <sub>IN_high</sub>		2.1	-	-	V
Input Current – High	I <sub>IN_high</sub>	V <sub>IN</sub> = 2.1 V	-	-	10	μΑ
Input Hysteresis Voltage	V <sub>IN_hyst</sub>		-	0.2	-	V
Input Clamp Voltage	V <sub>IN_cl</sub>	I <sub>IN</sub> = 1 mA	12	13	14	V
		I <sub>IN</sub> = -1 mA	-14	-13	-12	
CS_EN Voltage - Low	V <sub>CSE_low</sub>		-	-	0.9	V
CS_EN Current – Low	I <sub>CSE_low</sub>	V <sub>CS_EN</sub> = 0.9 V	1	-	-	μΑ
CS_EN Voltage - High	V <sub>CSE_high</sub>		2.1	-	-	V
CS_EN Current – High	I <sub>CSE_high</sub>	V <sub>CS_EN</sub> = 2.1 V	-	-	10	μΑ
CS_EN Hysteresis Voltage	V <sub>CSE_hyst</sub>		-	0.2	-	V
CS_EN Clamp Voltage	V <sub>CSE_cl</sub>	I <sub>CS_EN</sub> = 1 mA	12	13	14	V
		I <sub>CS_EN</sub> = -1 mA	-14	-13	-12	1

#### Table 7. SWITCHING CHARACTERISTICS (V\_D = 13 V, -40°C $\leq$ T\_J $\leq$ 150°C)

				Value		
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Turn-On Delay Time	t <sub>d_on</sub>	$V_{IN}$ high to 20% $V_{OUT}\!,R_L$ = 13 $\Omega\!,T_J$ = 25°C	5	70	120	μs
Turn-Off Delay Time	t <sub>d_off</sub>	$V_{IN}$ low to 80% $V_{OUT}\!,R_L$ = 13 $\Omega,T_J$ = 25°C	5	40	100	μs
Slew Rate On	dV <sub>out</sub> /dt <sub>on</sub>	20% to 80% V_OUT, R_L = 13 $\Omega,T_J$ = 25°C	0.1	0.27	0.7	V/μs
Slew Rate Off	dV <sub>out</sub> /dt <sub>off</sub>	80% to 20% V <sub>OUT</sub> , R <sub>L</sub> = 13 $\Omega$ , T <sub>J</sub> = 25°C	0.1	0.35	0.7	V / μs
Turn–On Switching Loss (Note 7)	E <sub>on</sub>	R <sub>L</sub> = 13 Ω	-	0.15	0.18	mJ
Turn–Off Switching Loss (Note 7)	E <sub>off</sub>	R <sub>L</sub> = 13 Ω	-	0.1	0.18	mJ
Differential Pulse Skew, (t <sub>(OFF)</sub> – t <sub>(ON)</sub> ) see Figure 4 (Switching Characteristics)	t <sub>skew</sub>	R <sub>L</sub> = 13 Ω	-50	-	50	μs

7. Not subjected to production testing.

#### Table 8. OUTPUT DIODE CHARACTERISTICS

			Value			
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Forward Voltage	V <sub>F</sub>	$I_{OUT} = -1 \text{ A}, \text{ T}_{J} = 150^{\circ}\text{C}$	1	-	0.7	V

#### Table 9. PROTECTION FUNCTIONS (Note 8) (7 V $\leq$ V\_D $\leq$ 18 V, -40°C $\leq$ T\_J $\leq$ 150°C)

			Value			
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Temperature Shutdown (Note 9)	T <sub>SD</sub>		150	175	200	°C
Temperature Shutdown Hysteresis (T <sub>SD</sub> – T <sub>R</sub> ) (Note 9)	T <sub>SD_hyst</sub>		-	7	-	°C
Reset Temperature (Note 9)	T <sub>R</sub>		T <sub>RS</sub> +1	T <sub>RS</sub> +7	-	°C
Thermal Reset of Status (Note 9)	T <sub>RS</sub>		135	-	-	°C
Delta T Temperature Limit (Note 9)	T <sub>DELTA</sub>	$T_{J} = -40^{\circ}C, V_{D} = 13 V$	-	60	-	°C
DC Output Current Limit	I <sub>limH</sub>	V <sub>D</sub> = 13 V	8	12	16	А
		4 V < V <sub>D</sub> < 18 V	-	-	16	А
Short Circuit Current Limit during Thermal Cycling (Note 9)	IlimTCycling	V <sub>D</sub> = 13 V T <sub>R</sub> < Tj < T <sub>TSD</sub>	-	4	-	А
Switch Off Output Clamp Voltage	V <sub>OUT_clamp</sub>	$I_{OUT} = 0.2 \text{ A}, V_{IN} = 0 \text{ V}, \text{ L} = 20 \text{ mH}$	V <sub>D</sub> – 41	V <sub>D</sub> - 46	V <sub>D</sub> - 52	V
Overvoltage Protection	V <sub>OV</sub>	V <sub>IN</sub> = 0 V, I <sub>D</sub> = 20 mA	41	46	52	V
Output Voltage Drop Limitation	V <sub>DS_ON</sub>	I <sub>OUT</sub> = 0.07 A	-	20	-	mV

 To ensure long term reliability during overload or short circuit conditions, protection and related diagnostic signals must be used together with a fitting hardware & software strategy. If the device operates under abnormal conditions, this hardware & software solution must limit the duration and number of activation cycles.

9. Not subjected to production testing.

#### Table 10. OPEN–LOAD DETECTION (7 V $\leq$ V\_D $\leq$ 18 V, –40°C $\leq$ T\_J $\leq$ 150°C)

			Value			
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Open-load Off State Detection Threshold	V <sub>OL</sub>	$V_{IN}$ = 0 V, $V_{CS}$ _EN = 5 V	2	-	4	V
Open-load Detection Delay at Turn Off	<sup>t</sup> d_OL_off		100	350	700	μs
Off State Output Current	I <sub>OLOFF1</sub>	V <sub>IN</sub> = 0 V, V <sub>OUT</sub> = V <sub>OL</sub>	-3	-	3	μA
Output rising edge to CS rising edge during open load	t <sub>d_OL</sub>	$V_{OUT}$ = 4 V, $V_{IN}$ = 0 V $V_{CS}$ = 90% of $V_{CS}$ High	-	5	30	μs

#### Table 11. CURRENT SENSE CHARACTERISTICS (7 V $\leq$ V\_D $\leq$ 18 V, -40°C $\leq$ T\_J $\leq$ 150°C)

			Value			
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Current Sense Ratio	K <sub>0</sub>	$I_{OUT}$ = 0.010 A, $V_{CS}$ = 0.5 V, $V_{CS}$ _EN = 5 V	260	_	800	
Current Sense Ratio	K <sub>1</sub>	$I_{OUT}$ = 0.025 A, $V_{CS}$ = 0.5 V, $V_{CS}$ _EN = 5 V	265	490	720	
Current Sense Ratio Drift (Note 11)	$\Delta K_1 / K_1$	$I_{OUT}$ = 0.025 A, $V_{CS}$ = 0.5 V, $V_{CS}$ _EN = 5 V	-25	-	25	%
Current Sense Ratio	K <sub>2</sub>	$I_{OUT} = 0.07 \text{ A}, V_{CS} = 4 \text{ V}, V_{CS}_{EN} = 5 \text{ V}$	270	475	675	
Current Sense Ratio Drift (Note 11)	$\Delta K_2 / K_2$	I <sub>OUT</sub> = 0.07 A, V <sub>CS_EN</sub> = 5 V	-20	_	20	%
Current Sense Ratio	K <sub>3</sub>	$I_{OUT} = 0.15 \text{ A}, V_{CS} = 4 \text{V}, V_{CS}_{EN} = 5 \text{ V}$	275	475	625	
Current Sense Ratio Drift (Note 11)	$\Delta K_3 / K_3$	I <sub>OUT</sub> = 0.15 A, V <sub>CS_EN</sub> = 5 V	-15	_	15	%
Current Sense Ratio	K <sub>4</sub>	$I_{OUT} = 0.7 \text{ A}, V_{CS} = 4 \text{ V}, V_{CS}_{EN} = 5 \text{ V}$	375	475	540	
Current Sense Ratio Drift (Note 11)	$\Delta K_4 / K_4$	I <sub>OUT</sub> = 0.7 A, V <sub>CS_EN</sub> = 5 V	-10	-	10	%
Current Sense Ratio	K <sub>5</sub>	$I_{OUT} = 2 \text{ A}, V_{CS} = 4 \text{ V}, V_{CS}_{EN} = 5 \text{ V}$	420	450	480	
Current Sense Ratio Drift (Note 11)	$\Delta K_5 / K_5$	I <sub>OUT</sub> = 2 A, V <sub>CS_EN</sub> = 5 V	-5	-	5	%
Current Sense Leakage Current	CS <sub>Ilkg</sub>	$I_{OUT} = 0 \text{ A}, V_{CS} = 0 \text{ V}$ $V_{CS_{EN}} = 5 \text{ V}, V_{IN} = 0 \text{ V}$	-	-	1	μA
		$I_{OUT} = 0 \text{ A}, V_{CS} = 0 \text{ V}$ $V_{CS\_EN} = 5 \text{ V}, V_{IN} = 5 \text{ V}$	-	-	2	
		$I_{OUT} = 1 \text{ A}, V_{CS} = 0 \text{ V}$ $V_{CS\_EN} = 0 \text{ V}, V_{IN} = 5 \text{ V},$	-	-	0.5	
CS Max Voltage	CS <sub>Max</sub>	$V_D$ = 7 V, $V_{IN}$ = 5 V, $R_{CS}$ = 2.7 k $\Omega$ , $I_{OUT}$ = 2 A, $T_J$ = 150°C, $V_{CS}$ _EN = 5 V	5	-	7	V
Current Sense Voltage in Fault Con- dition (Note 10)	V <sub>CS_fault</sub>	$V_{D}$ = 13 V, $V_{IN}$ = 0 V, $R_{CS}$ = 1 k, $V_{OUT}$ = 4 V, $V_{CS}$ _EN = 5 V	-	10	-	V
Current Sense Current in Fault Con- dition (Note 10)	I <sub>CS_fault</sub>	$V_{D} = 13 \text{ V}, V_{CS} = 5 \text{ V}, V_{IN} = 0 \text{ V}, V_{OUT} = 4 \text{ V}, V_{CS_EN} = 5 \text{ V}$	7	20	30	mA
Output Saturation Current (Note 10)	I <sub>OUT_sat</sub>	$V_D = 7 V, V_{CS} = 4 V, V_{IN} = 5 V, T_J = 150^{\circ}C, V_{CS_EN} = 5 V$	2	-	-	A
CS_EN High to CS High Delay Time	t <sub>CS_High1</sub>		-	-	100	μs
CS_EN Low to CS Low Delay Time	t <sub>CS_Low1</sub>		-	5	25	μs
V <sub>in</sub> High to CS High Delay Time	t <sub>CS_High2</sub>			100	250	μs
V <sub>in</sub> Low to CS Low Delay Time	t <sub>CS_Low2</sub>	$V_{IN}$ = 5 to 0 V, $V_{CS\_EN}$ = 5 V, $R_{CS}$ = 1 k $\Omega$ , $R_L$ = 13 $\Omega$	-	50	250	μs
Delay Time I <sub>D</sub> Rising Edge to Rising Edge of CS	$\Delta t_{CS_High2}$	$V_{IN}$ = 5 V, $V_{CS-EN}$ = 5 V R <sub>CS</sub> = 1 k $\Omega$ , I <sub>CS</sub> = 90% of I <sub>CS</sub> Max	-	-	100	μs

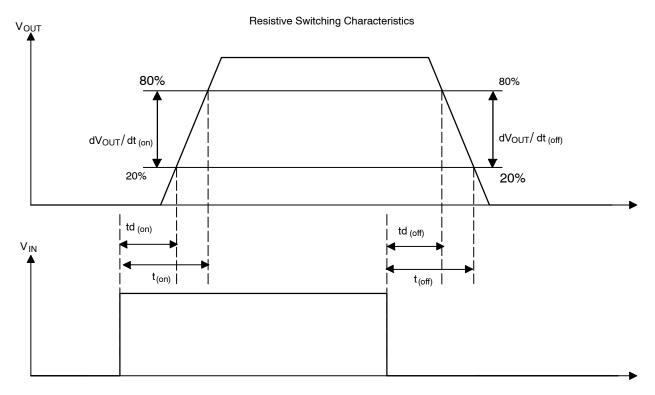
10. The following fault conditions included are: Over-temperature, Power Limitation, and OFF State Open-Load Detection. 11. Not subjected to production testing.

#### Table 12. TRUTH TABLE

Conditions	Input	Output	CS (V <sub>CS_EN</sub> = 5 V) (Note 12)
Normal Operation	L H	L H	0 I <sub>CS</sub> = I <sub>OUT</sub> /K <sub>NOMINAL</sub>
Overtemperature	L H	L	0 V <sub>CS_fault</sub>
Undervoltage	L H	L	0 0
Overload	H H	H (no active current mgmt) Cycling (active current mgmt)	I <sub>CS</sub> = I <sub>OUT</sub> /K <sub>NOMINAL</sub> V <sub>CS_fault</sub>
Short circuit to Ground	L H	L	0 V <sub>CS_fault</sub>
OFF State Open Load	L	н	V <sub>CS_fault</sub>

12. If V<sub>CS\_EN</sub> is low, the Current Sense output is at a high impedance, its potential depends on leakage currents and external circuitry.

#### WAVEFORMS AND GRAPHS





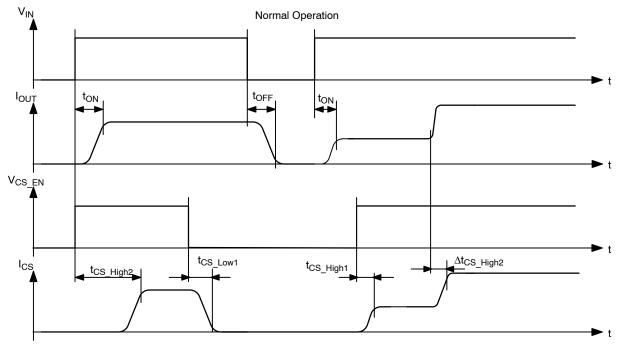
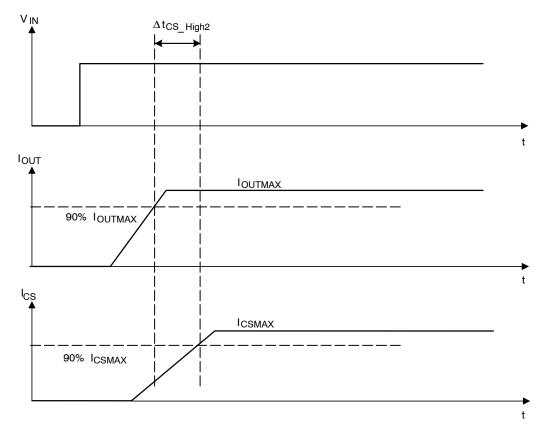
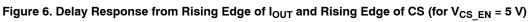


Figure 5. Normal Operation with Current Sense Timing Characteristics





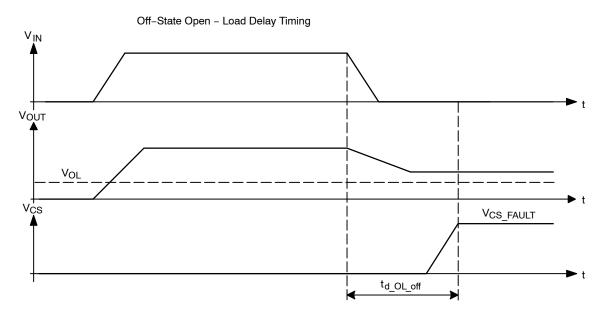


Figure 7. OFF-State Open-Load Flag Delay Timing

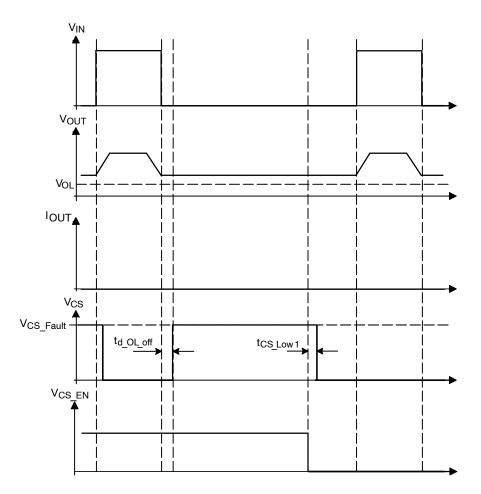


Figure 8. Off-State Open-Load with Added External Components

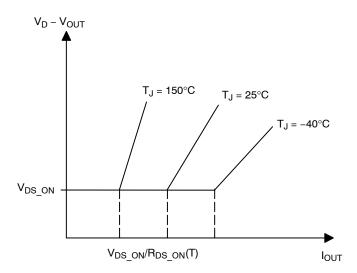


Figure 9. Voltage Drop Limitation for  $V_{\text{DS}\_\text{ON}}$ 

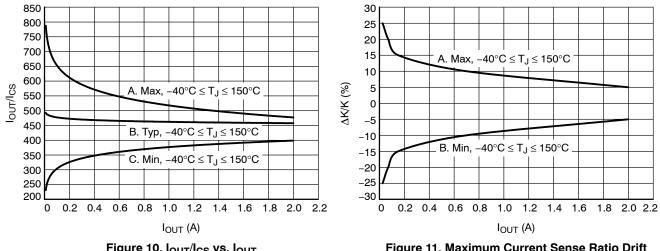




Figure 11. Maximum Current Sense Ratio Drift vs. Load Current

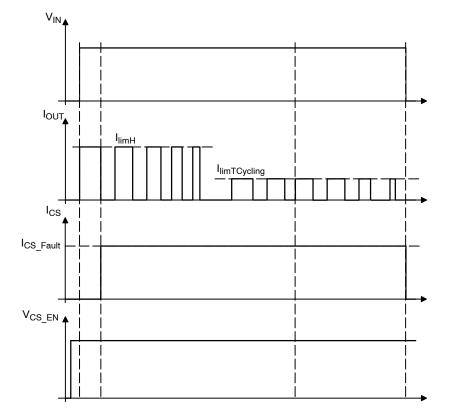


Figure 12. Short to GND or Overload

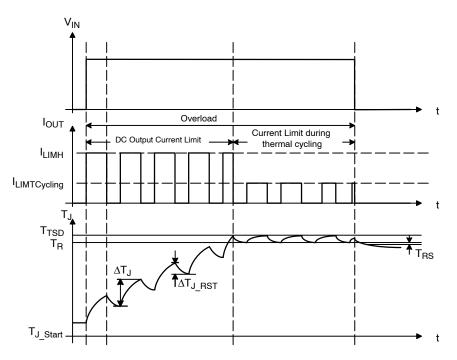
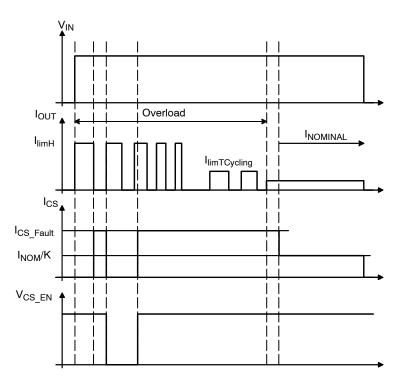
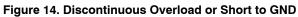
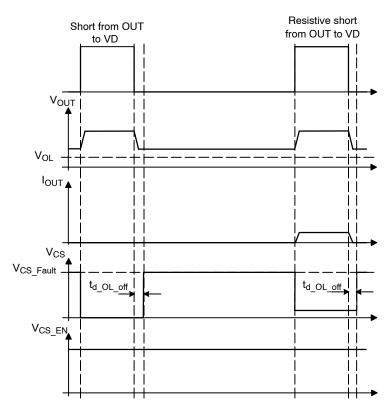


Figure 13. How  $\rm T_J$  progresses During Short to GND or Overload









### **TYPICAL CHARACTERISTICS**

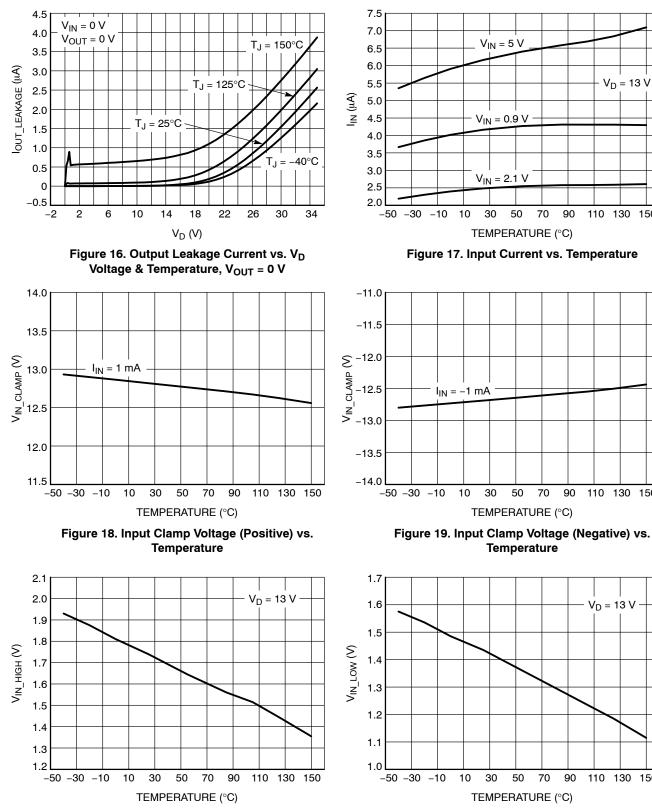


Figure 21. V<sub>IN</sub> Threshold Low vs. Temperature

V<sub>D</sub> = 13 V

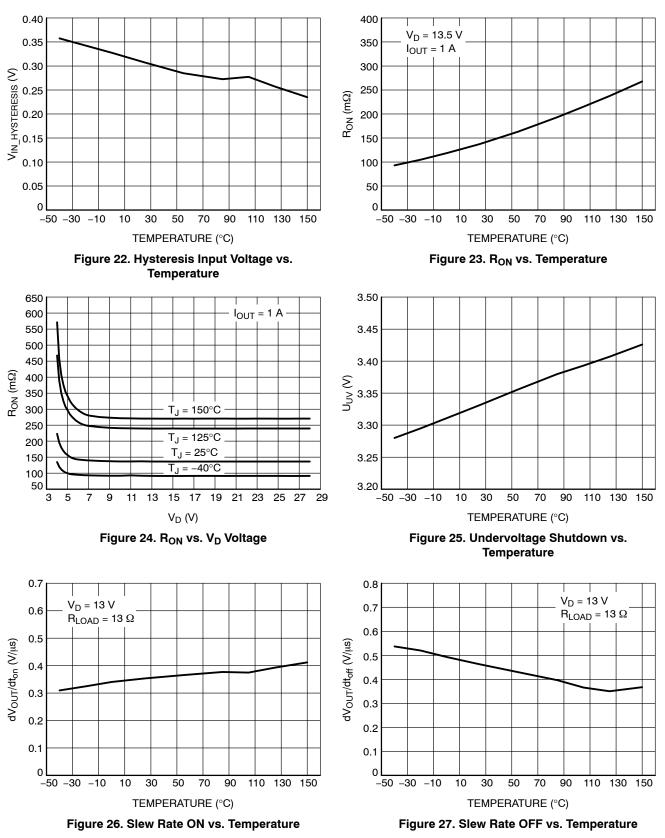
110 130 150

110 130 150

V<sub>D</sub> = 13 V

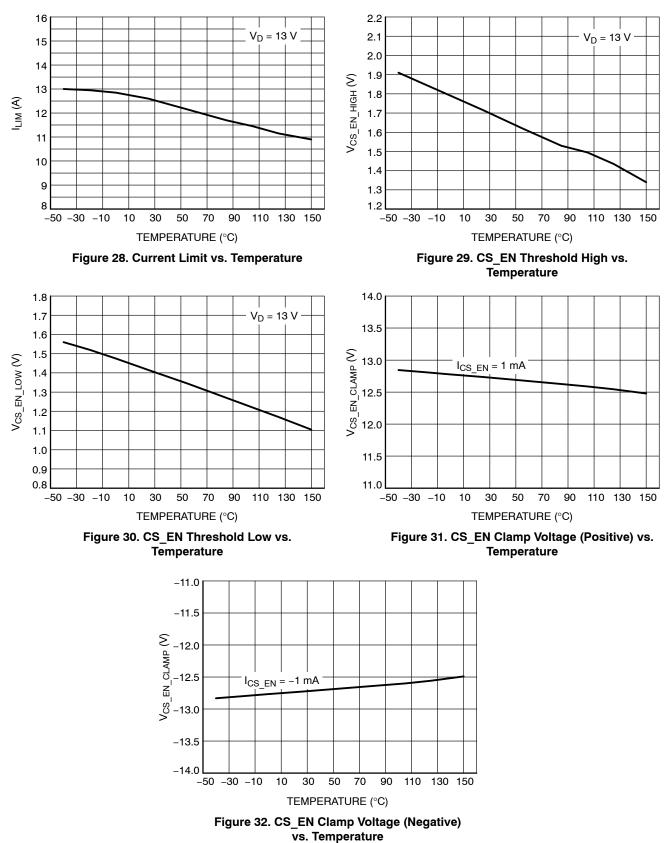
110 130 150

Figure 20. V<sub>IN</sub> Threshold High vs. Temperature



#### **TYPICAL CHARACTERISTICS**





ISO 7637-2:2011	Test Severity Levels					
Test Pulse			Delays and Impedance	# of Pulses or Test Time	Pulse / Burst Rep. Time	
1	-112	-150	2 ms, 10 Ω	500 pulses	0.5 s	
2a	55	112	0.05 ms, 2 $\Omega$	500 pulses	0.5 s	
3a	-165	-220	0.1 μs, 50 Ω	1 h	100 ms	
3b	112	150	0.1 μs, 50 Ω	1 h	100 ms	
ISO 7637-2:2011	Test R	esults				
Test Pulse	ш	IV				
1		А				
2a	С	E				
3a		А				
3b		А				
Class			Function	nal Status		
А	All functions of a device perform as designed during and after exposure to disturbance.					
В	All functions of a device perform as designed during exposure. However, one or more of them can go beyond speci- fied tolerance. All functions return automatically to within normal limits after exposure is removed. Memory functions shall remain class A.					
С	One or more functions of a device do not perform as designed during exposure but return automatically to normal operation after exposure is removed.					
D	One or more functions of a device do not perform as designed during exposure and do not return to normal operation until exposure is removed and the device is reset by simple "operator/use" action.					
Е	One or more functions of a device do not perform as designed during and after exposure and cannot be returned to proper operation without replacing the device.					

### Table 13. ISO 7637-2: 2011(E) PULSE TEST RESULTS

## **APPLICATION INFORMATION**

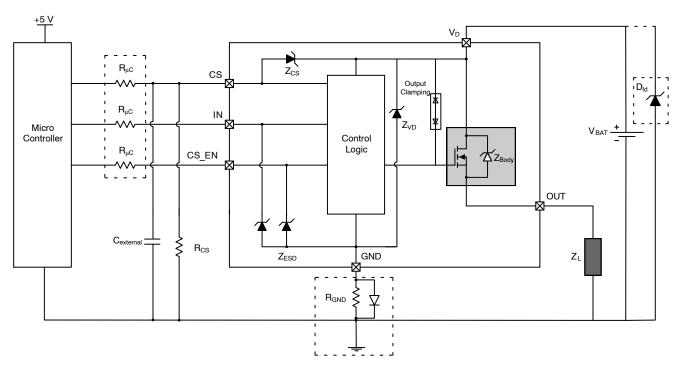


Figure 33. Application Schematic

#### Loss of Ground Protection

When device or ECU ground connection is lost and load is still connected to ground, the device will turn the output OFF. In loss of ground state, the output stage is held OFF independent of the state of the input. Input resistors are recommended between the device and microcontroller.

#### **Reverse Battery Protection**

## Solution 1: Resistor in the GND line only (no parallel Diode)

The following calculations are true for any type of load. In the case for no diode in parallel with  $R_{GND}$ , the calculations below explain how to size the resistor.

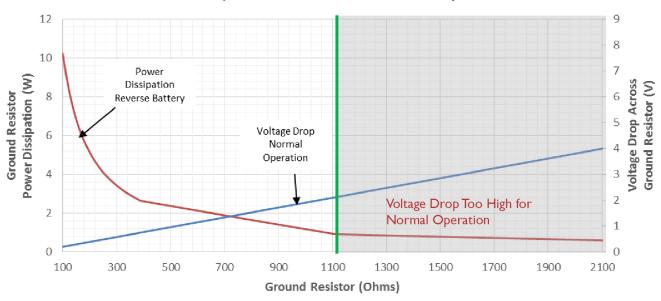
Consider the following parameters:

 $-I_{GND}$  Maximum = 200 mA for up to  $-V_D$  = 32 V.

Where  $-I_{GND}$  is the DC reverse current through the GND pin and  $-V_D$  is the DC reverse battery voltage.

$$-I_{GND} = \frac{-V_D}{R_{GND}}$$
 (eq. 1)

Since this resistor can be used amongst multiple High–Side devices, please take note the sum of the maximum active GND currents ( $I_{GND(On)max}$ ) for each device when sizing the resistor. Please note that if the microprocessor GND is not shared by the device GND, then  $R_{GND}$  produces a shift of ( $I_{GND(On)max} \times R_{GND}$ ) in the input thresholds and CS output values. If the calculated power dissipation leads to too large of a resistor size or several devices have to share the same resistor, please look at the second solution for Reverse Battery Protection. Refer to Figure 35 for selecting the proper  $R_{GND}$ .



#### Normal Operation VIN = 5 V, Reverse Battery = 32 V

Figure 34. Reverse Battery R<sub>GND</sub> Considerations

# Solution 2: Diode $(D_{GND})$ in parallel with RGND in the ground line.

A resistor value of  $R_{GND} = 1$  kOhm should be selected and placed in parallel to  $D_{GND}$  if the device drives an inductive load. The diode ( $D_{GND}$ ) provides a ~600–700 mV shift in the input threshold and current sense values if the micro controller ground is not common to the device ground. This shift will not vary even in the case of multiple high-side devices using the same resistor/diode network.

#### **Undervoltage Protection**

The device has two under-voltage threshold levels,  $V_{D\_MIN}$  and  $V_{UV}$ . Switching function (ON/OFF) requires supply voltage to be at least  $V_{D\_MIN}$ . The device features a lower supply threshold  $V_{UV}$ , above which the output can

remain in ON state. While all protection functions are guaranteed when the switch is ON, diagnostic functions are operational only within nominal supply voltage range  $V_{D.}$ 

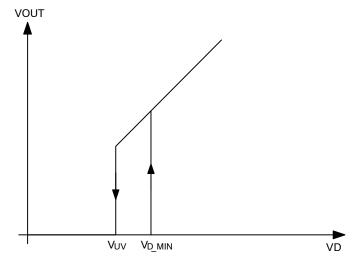


Figure 35. Undervoltage Behavior

#### **Overvoltage Protection**

The NCV84140 has two Zener diodes Z<sub>VD</sub> and Z<sub>CS</sub>, which provide integrated overvoltage protection. Z<sub>VD</sub> protects the logic block by clamping the voltage between supply pin  $V_D$  and ground pin GND to  $V_{ZVD}$ . Z<sub>CS</sub> limits voltage at current sense pin CS to  $V_D - V_{ZCS}$ . The output power MOSFET's output clamping diodes provide protection by clamping the voltage across the MOSFET (between V<sub>D</sub> pin and OUT pin) to V<sub>CLAMP</sub>. During overvoltage protection, current flowing through Z<sub>VD</sub>, Z<sub>CS</sub> and the output clamp must be limited. Load impedance ZL limits the current in the body diode ZBody. In order to limit the current in  $Z_{VD}$  a resistor,  $R_{GND}$  (150  $\Omega$ ), is required in the GND path. External resistors R<sub>CS</sub> and R<sub>SENSE</sub> limit the current flowing through Z<sub>CS</sub> and out of the CS pin into the micro-controller I/O pin. With RGND, the GND pin voltage is elevated to  $V_D - V_{ZVD}$  when the supply voltage  $V_D$  rises above V<sub>ZVD</sub>. ESD diodes Z<sub>ESD</sub> pull up the voltage at logic pins IN, CS\_EN close to the GND pin voltage  $V_D - V_{ZVD}$ . External resistors RIN, and RCS EN are required to limit the current flowing out of the logic pins into the micro-controller I/O pins. During overvoltage exposure, the device transitions into a self-protection state, with

automatic recovery after the supply voltage comes back to the normal operating range. The specified parameters as well as short circuit robustness and energy capability cannot be guaranteed during overvoltage exposure.

#### **Overload Protection**

Current limitation as well as overtemperature shutdown mechanisms are integrated into NCV84140 to provide protection from overload conditions such as bulb inrush or short to ground.

#### **Current Limitation**

In case of overload, NCV84140 limits the current in the output power MOSFET to a safe value. Due to high power dissipation during current limitation, the device's junction temperature increases rapidly. In order to protect the device, the output driver is shut down by one of the two overtemperature protection mechanisms. The output current limit is dependent on the device temperature, and will fold back once the die reaches thermal shutdown. If the input remains active during the shutdown, the output power MOSFET will automatically be re-activated after a minimum OFF time or when the junction temperature returns to a safe level.

#### **Output Clamping with Inductive Load Switch Off**

The output voltage Vour drops below GND potential when switching off inductive loads. This is because the inductance develops a negative voltage across the load in response to a decaying current. The integrated clamp of the device clamps the negative output voltage to a certain level relative to the supply voltage VBAT. During output clamping with inductive load switch off, the energy stored in the inductance is rapidly dissipated in the device resulting in high power dissipation. This is a stressful condition for the device and the maximum energy allowed for a given load inductance should not be exceeded in any application.

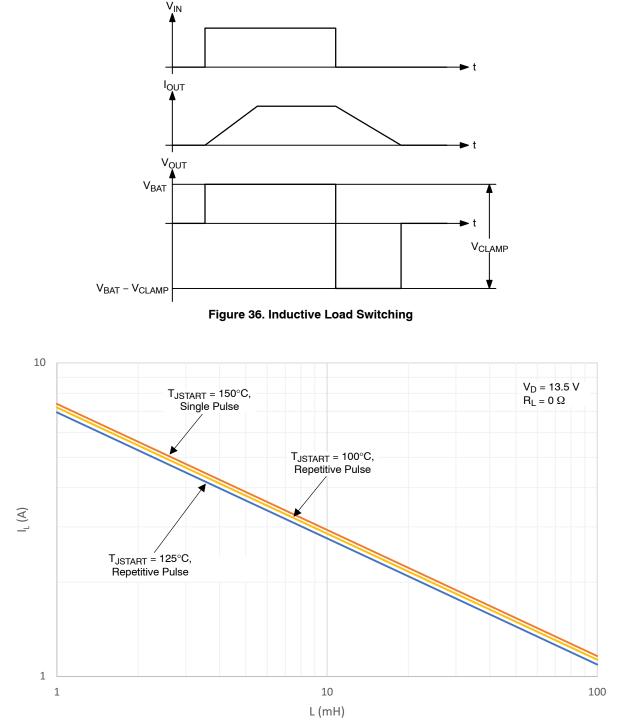


Figure 37. Maximum Switch–Off Current vs. Load Inductance, V<sub>D</sub> = 13.5 V, R<sub>L</sub> = 0  $\Omega$ 

#### **Open Load Detection in OFF State**

Open load diagnosis in OFF state can be performed by activating an external resistive pull-up path ( $R_{PU}$ ) to  $V_{BAT}$ . To calculate the pull-up resistance, external leakage

currents (designed pull-down resistance, humidity-induced leakage etc) as well as the open load threshold voltage  $V_{OL}$  have to be taken into account.

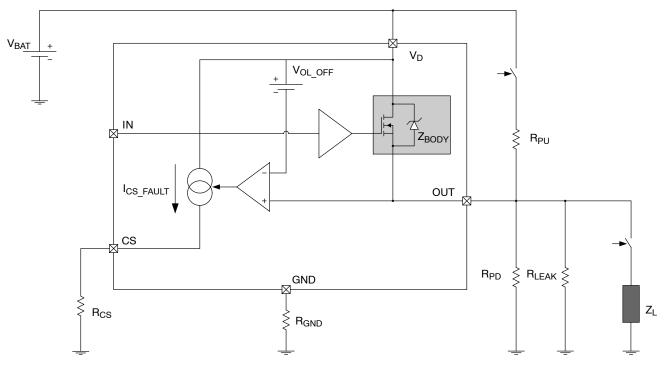


Figure 38. Open Load Detection in Off State

#### **Current Sense in PWM Mode**

When operating in PWM mode, the current sense functionality can be used, but the timing of the input signal and the response time of the current sense need to be considered. When operating in PWM mode, the following performance is to be expected. The CS\_EN pin should be held high to eliminate any unnecessary delay time to the circuit. When  $V_{IN}$  switches from low to high, there will be a typical delay (t<sub>CS\_High2</sub>) before the current sense responds. Once this timing delay has passed, the rise time of the current

sense output ( $\Delta t_{CS\_High2}$ ) also needs to be considered. When  $V_{IN}$  switches from high to low a delay time ( $t_{CS\_Low1}$ ) needs to be considered. As long as these timing delays are allowed, the current sense pin can be operated in PWM mode.

#### **EMC** Performance

If better EMC performance is needed, connect a  $C_1 = 100 \text{ nF}$ ,  $C_2 = C_3 = 10 \text{ nF}$  ceramic capacitors to the pins as close to the device as possible according to Figure 39.

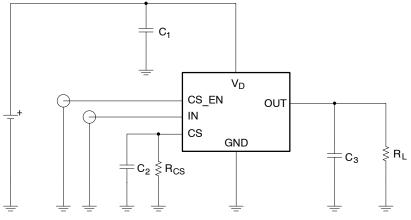
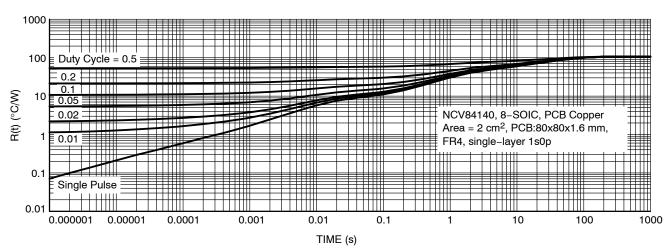


Figure 39. EMC Capacitors Placement



#### PACKAGE AND PCB THERMAL DATA



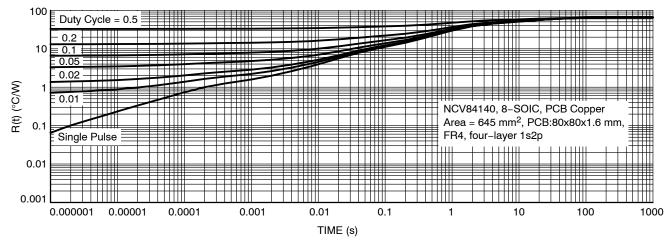


Figure 41. Junction to Ambient Transient Thermal Impedance (645 mm<sup>2</sup> Cu Area)





\*For additional information on our Pb–Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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STYLE 3: PIN 1. DRAIN, DIE #1 2. DRAIN, #1 3. DRAIN, #2 4. DRAIN, #2 5. GATE, #2 6. SOURCE, #2 7. GATE, #1 8. SOURCE, #1
STYLE 7: PIN 1. INPUT 2. EXTERNAL BYPASS 3. THIRD STAGE SOURCE 4. GROUND 5. DRAIN 6. GATE 3 7. SECOND STAGE Vd 8. FIRST STAGE Vd
STYLE 11: PIN 1. SOURCE 1 2. GATE 1 3. SOURCE 2 4. GATE 2 5. DRAIN 2 7. DRAIN 1 8. DRAIN 1
STYLE 15: PIN 1. ANODE 1 2. ANODE 1 3. ANODE 1 4. ANODE 1 5. CATHODE, COMMON 6. CATHODE, COMMON 7. CATHODE, COMMON 8. CATHODE, COMMON
STYLE 19: PIN 1. SOURCE 1 2. GATE 1 3. SOURCE 2 4. GATE 2 5. DRAIN 2 6. MIRROR 2 7. DRAIN 1 8. MIRROR 1
STYLE 23: PIN 1. LINE 1 IN 2. COMMON ANODE/GND 3. COMMON ANODE/GND 4. LINE 2 IN 5. LINE 2 OUT 6. COMMON ANODE/GND 7. COMMON ANODE/GND 8. LINE 1 OUT
STYLE 27: PIN 1. ILIMIT 2. OVLO 3. UVLO 4. INPUT+ 5. SOURCE 6. SOURCE 7. SOURCE 8. DRAIN

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